



APECE-302: Radio & Television Engineering

Applied Physics, Electronics & Communication Engineering

LEC PPT # 03
Modulation Techniques
(Part 1)



University of
Dhaka | APECE
DU

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Date: 2013 Year, 07 Month, 02 Day



Contents

- ❑ Introduction to Modulation
 - ❑ What, why and how!

- ❑ Analog Modulation/Demodulation: AM, FM, and PM
 - ❑ DSBSC, SSB, VSB

- ❑ Superheterodyne Receiver

- ❑ Digital Modulation/Demodulation: ASK/PSK/FSK/QAM/MSK/M-Ary

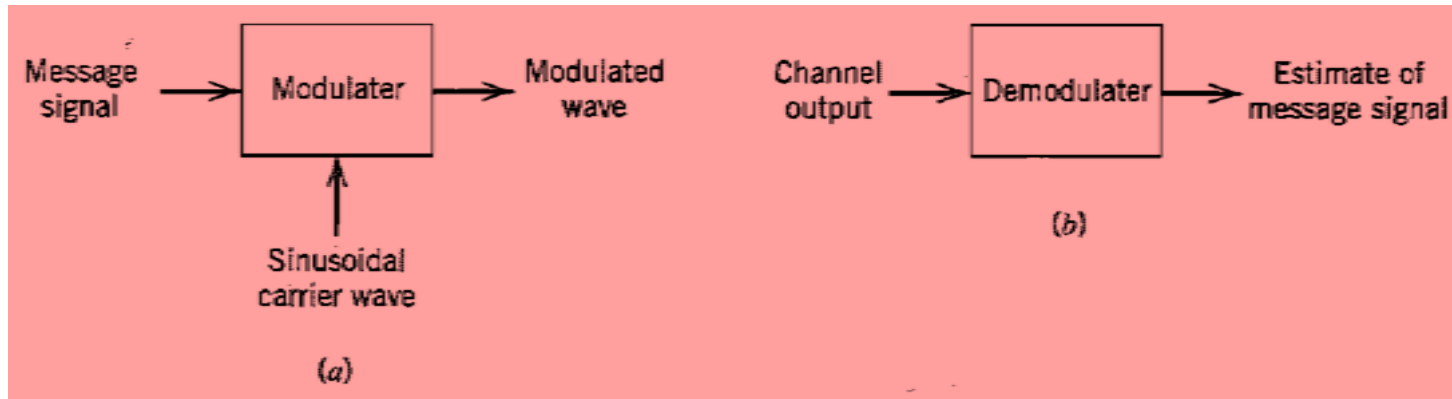
- ❑ Spread Spectrum Modulation: FHSS, DSSS

Introduction

- ❑ Information-bearing signals from Tx to Rx
 - ❑ Baseband signals
 - ❑ Modulating and modulated signal
 - ❑ Demodulation

- ❑ Reasons
 - ❑ Spectrum management
 - ❑ Different Stations locating
 - ❑ Antenna Height

Introduction



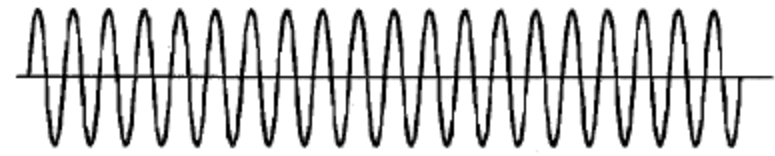
Introduction

- Continuous Modulation:
Amplitude and Angle
- Example: Amplitude and
frequency modulation at
RHS

AM



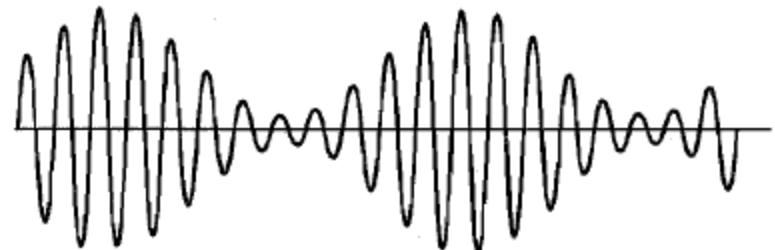
FM



(a)



(b)



(c)



(d)

Time →

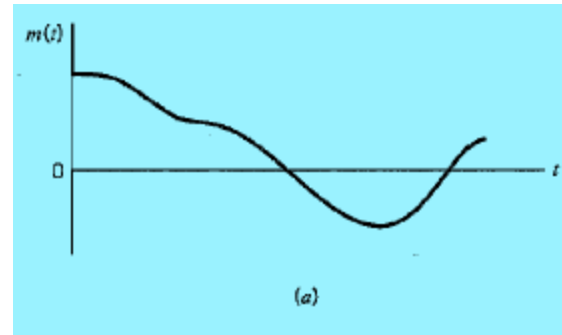
Amplitude Modulation

$$c(t) = A_c \cos(2\pi f_c t)$$

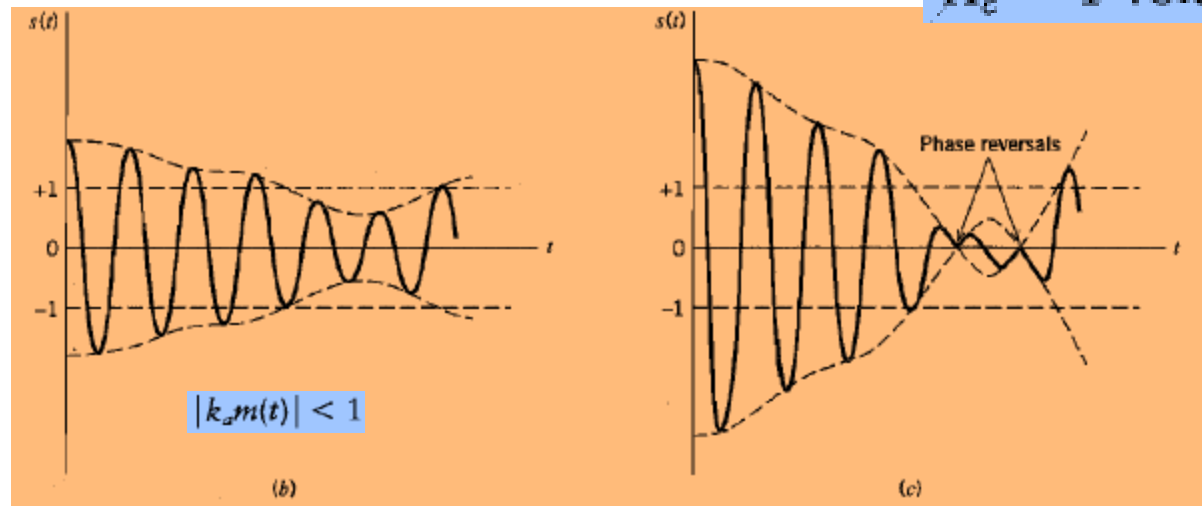
$$s(t) = A_c [1 + k_a m(t)] \cos(2\pi f_c t)$$



Amplitude sensitivity; ?unit



$$A_c = 1 \text{ volt.}$$



Conditions for AM

The amplitude of $k_a m(t)$ is always less than unity, that is,

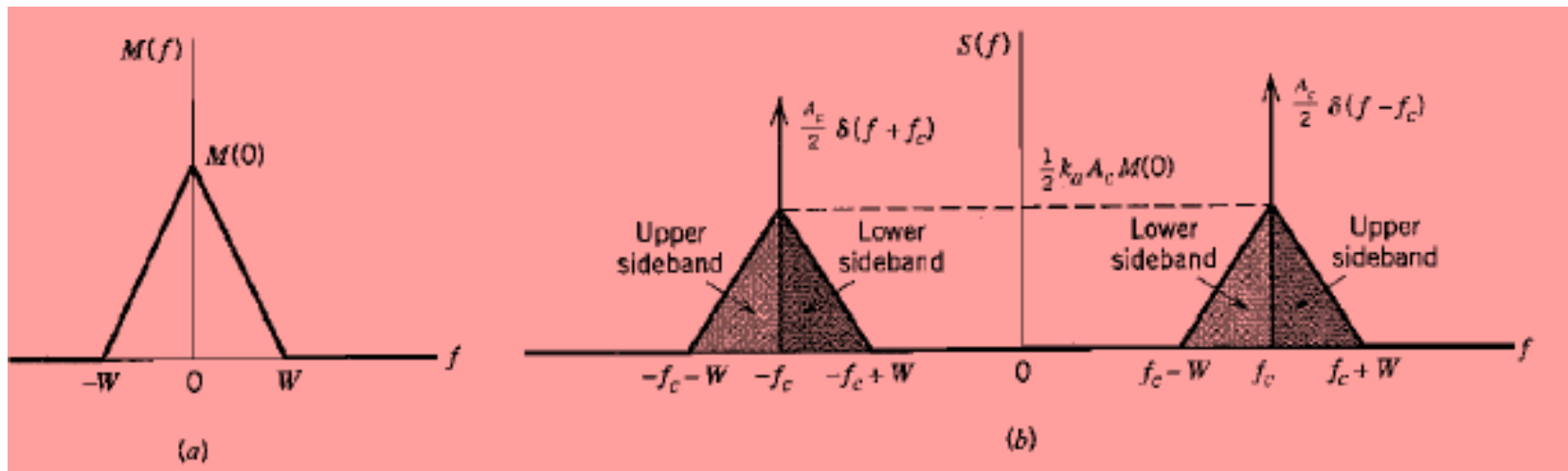
$$|k_a m(t)| < 1 \quad \text{for all } t$$

The carrier frequency f_c is much greater than the highest frequency component W of the message signal $m(t)$, that is

$$f_c \gg W \quad (2.4)$$

Spectrum Analysis in AM

$$S(f) = \frac{A_c}{2} [\delta(f - f_c) + \delta(f + f_c)] + \frac{k_a A_c}{2} [M(f - f_c) + M(f + f_c)]$$

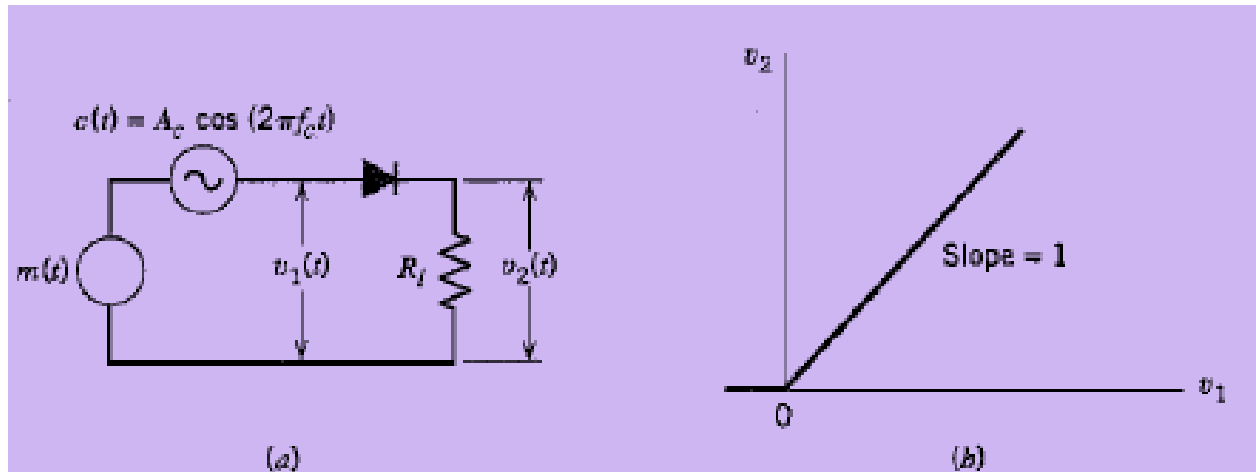


Spectrum Analysis of AM

- Realization of negative frequency!
- Must not be overlapped in LSB and USB
- Tx BW is twice the message BW

Virtues and Limitations of AM

- ❑ Oldest method and simplicity of implementation



Virtues and Limitations of AM

- ❑ Wasteful of power

- ❑ Wasteful of BW

- ❑ Overcome: DSB-SC, SSB w. system complexity
- ❑ ?Trade-off>>Linear modulation!

Single-tone Modulation

$$m(t) = A_m \cos(2\pi f_m t)$$

$$s(t) = A_c [1 + \mu \cos(2\pi f_m t)] \cos(2\pi f_c t)$$

Modulation Factor

$$\mu = \frac{A_{\max} - A_{\min}}{A_{\max} + A_{\min}}$$

$$s(t) = A_c \cos(2\pi f_c t) + \frac{1}{2} \mu A_c \cos[2\pi(f_c + f_m)t] + \frac{1}{2} \mu A_c \cos[2\pi(f_c - f_m)t]$$

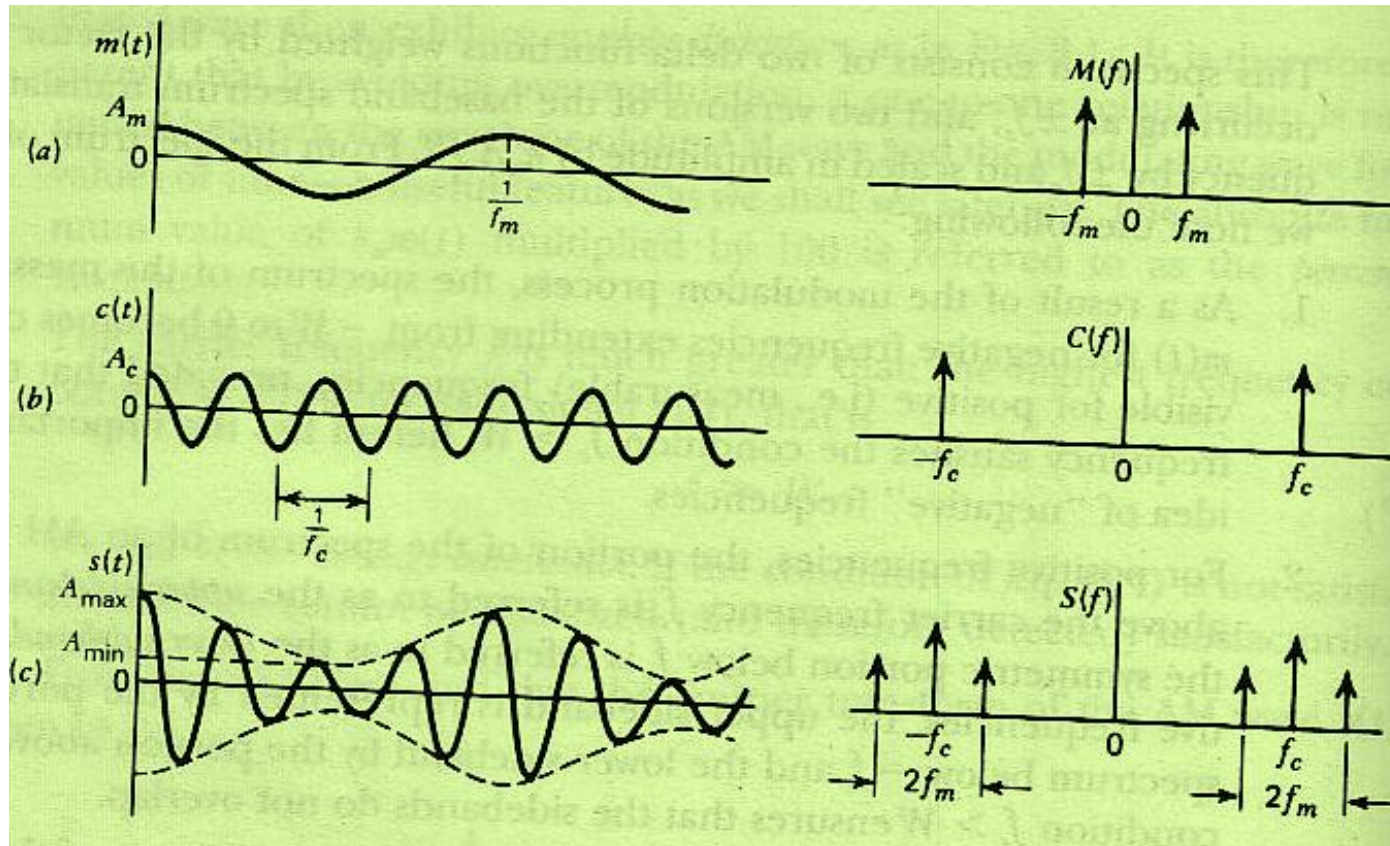
Single-tone Modulation

$$s(t) = A_c \cos(2\pi f_c t) + \frac{1}{2}\mu A_c \cos[2\pi(f_c + f_m)t] + \frac{1}{2}\mu A_c \cos[2\pi(f_c - f_m)t]$$



$$S(f) = \frac{1}{2}A_c[\delta(f - f_c) + \delta(f + f_c)] \\ + \frac{1}{4}\mu A_c[\delta(f - f_c - f_m) + \delta(f + f_c + f_m)] \\ + \frac{1}{4}\mu A_c[\delta(f - f_c + f_m) + \delta(f + f_c - f_m)]$$

Single-tone Modulation

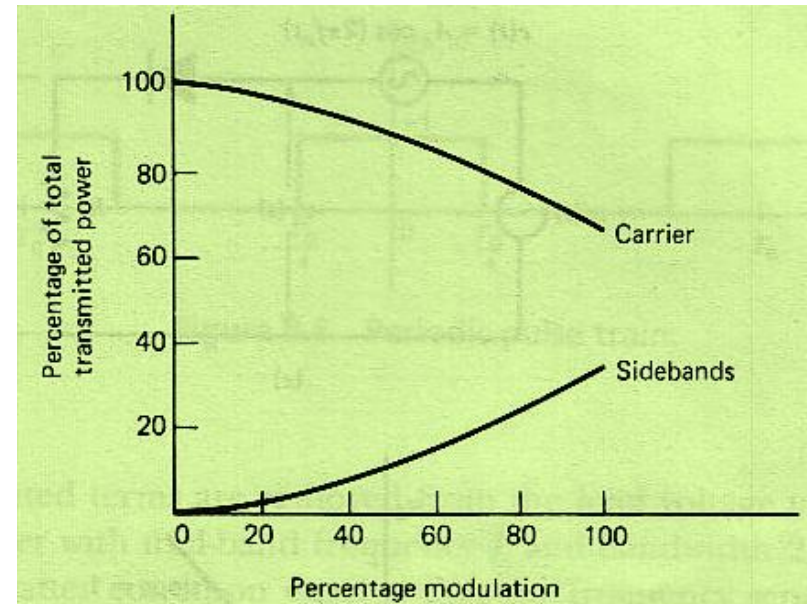


Single-tone Modulation

$$\text{Carrier power} = \frac{1}{2}A_c^2$$

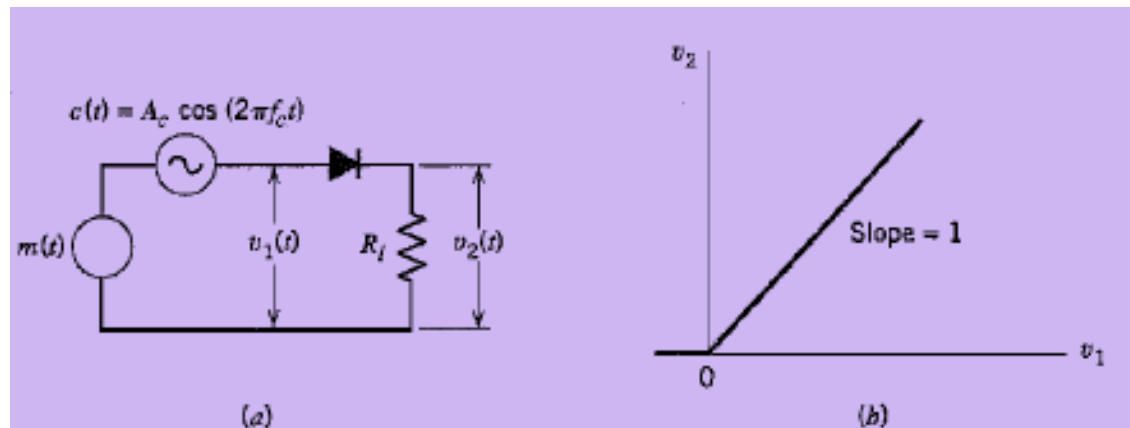
$$\text{Upper side-frequency power} = \frac{1}{8}\mu^2 A_c^2$$

$$\text{Lower side-frequency power} = \frac{1}{8}\mu^2 A_c^2$$



- $\mu \gg 1$ Ratio of total side-band power to the total power in the modulated wave
 - For 100% modulation

Switching Modulator



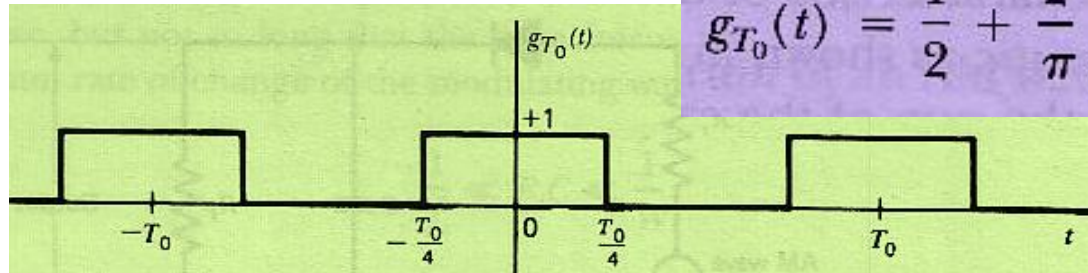
$$v_1(t) = A_c \cos(2\pi f_c t) + m(t)$$

where $|m(t)| \ll A_c$, the resulting load voltage $v_2(t)$ is

$$v_2(t) = \begin{cases} v_1(t), & c(t) > 0 \\ 0, & c(t) < 0 \end{cases}$$

Switching Modulator

$$v_2(t) \simeq [A_c \cos(2\pi f_c t) + m(t)] g_{T_0}(t)$$

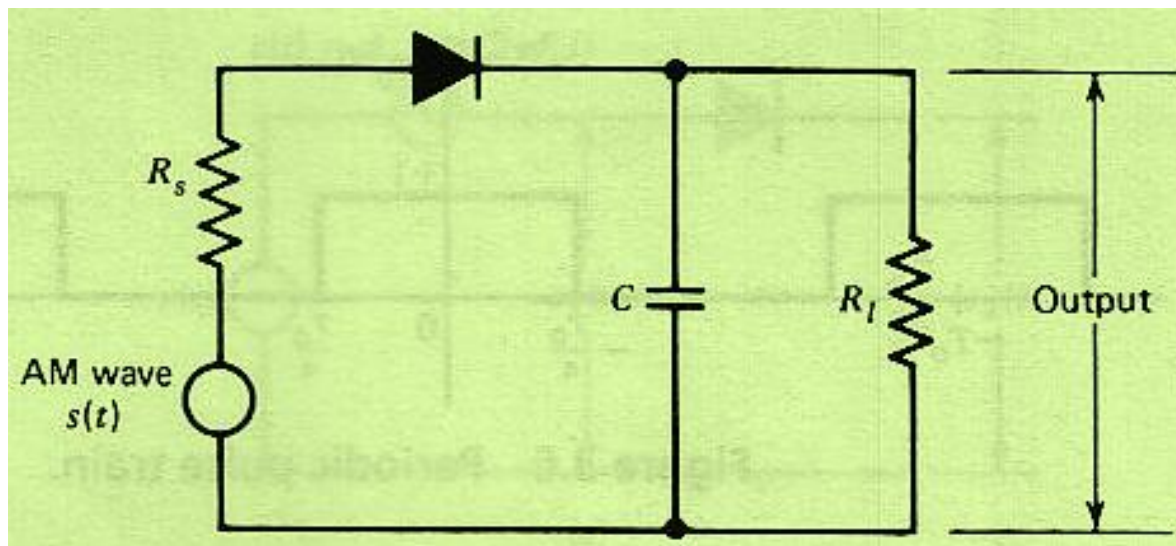


$$g_{T_0}(t) = \frac{1}{2} + \frac{2}{\pi} \sum_{n=1}^{\infty} \frac{(-1)^{n-1}}{2n-1} \cos[2\pi f_c t(2n-1)]$$

$$\frac{A_c}{2} \left[1 + \frac{4}{\pi A_c} m(t) \right] \cos(2\pi f_c t)$$

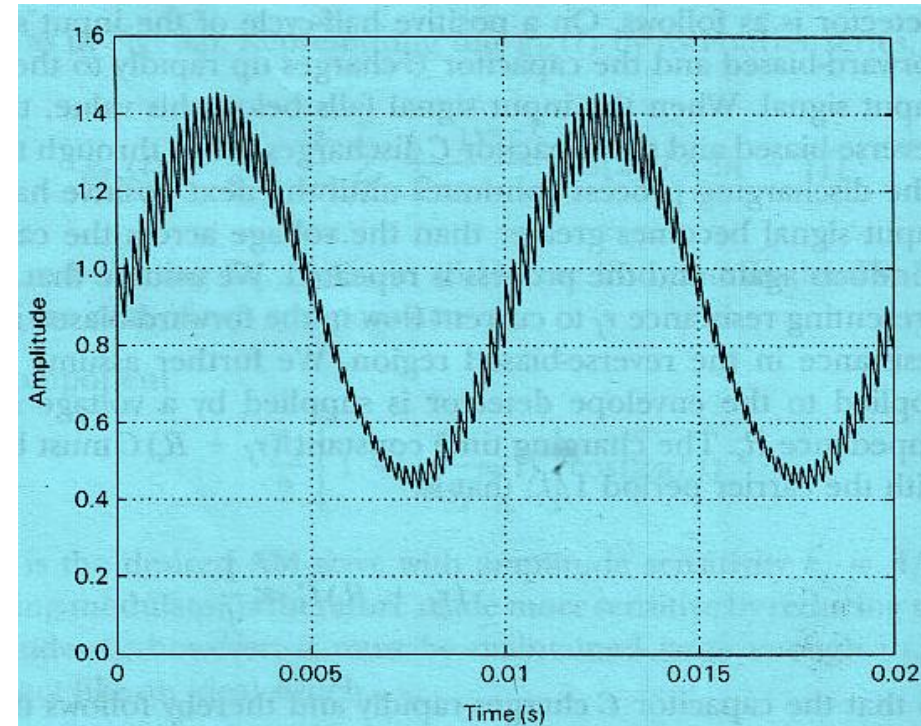
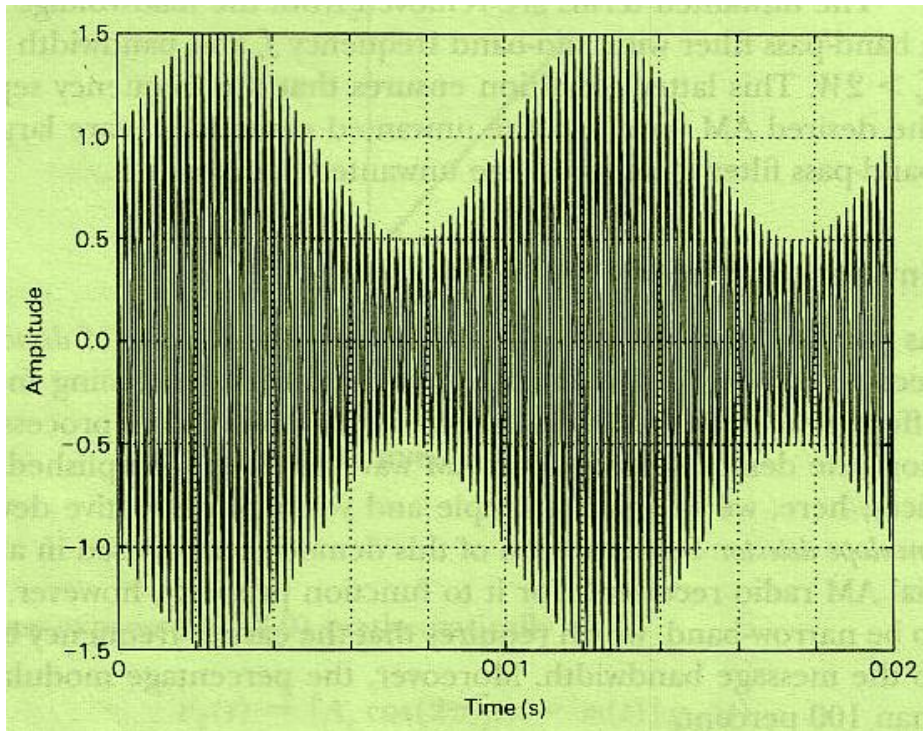
Unwanted components, the spectrum of which contains delta functions at $0, \pm 2f_c, \pm 4f_c$, and so on, and which occupy frequency intervals of width $2W$ centered at $0, \pm 3f_c, \pm 5f_c$, and so on, where W is the message bandwidth.

Envelope Detector



$$(r_f + R_s)C \ll \frac{1}{f_c}$$

Envelope Detector

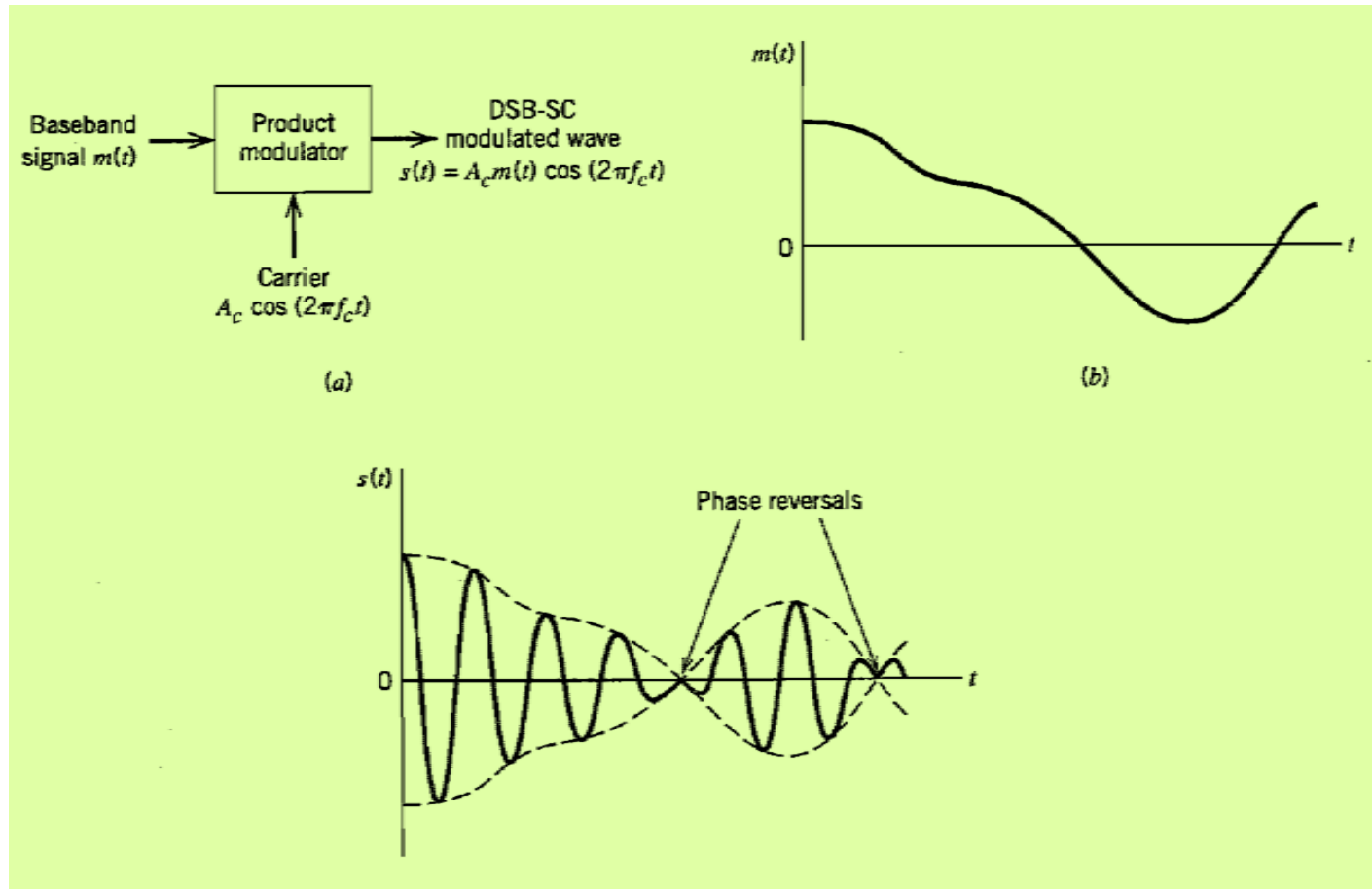


$$\frac{1}{f_c} \ll R_t C \ll \frac{1}{W}$$

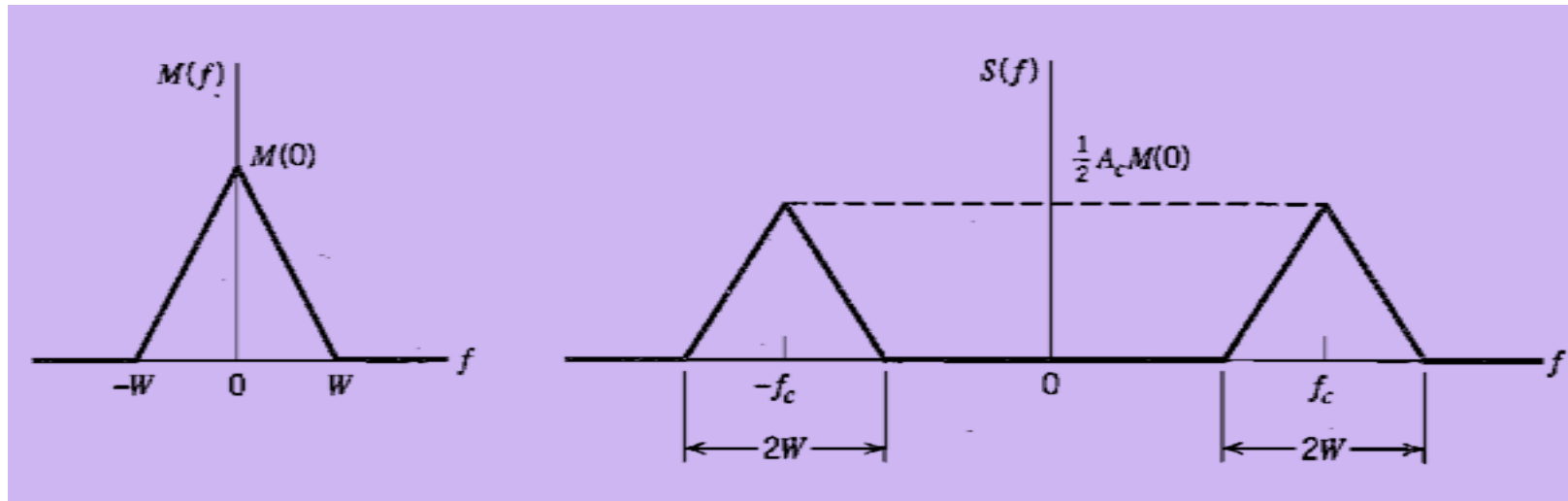
Envelope Detector

Source resistance	$R_s = 75 \Omega$
Forward resistance	$r_f = 25 \Omega$
Load resistance	$R_l = 10 k\Omega$
Capacitance	$C = 0.01 \mu F$
Message bandwidth	$W = 1 kHz$
Carrier frequency	$f_c = 20 kHz$

DSB-SC Modulation

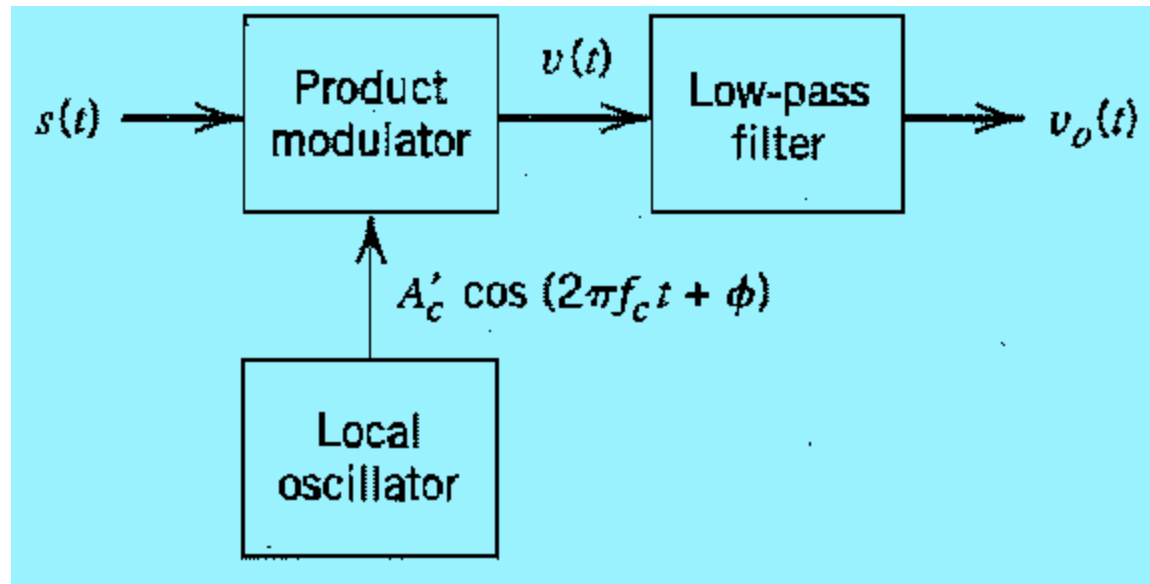


DSB-SC Modulation



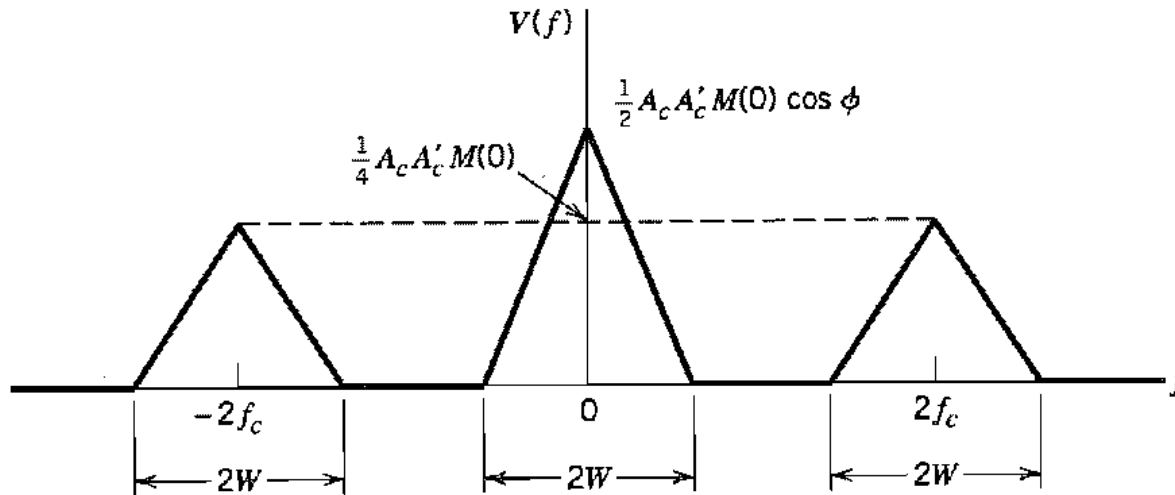
$$S(f) = \frac{1}{2} A_c [M(f - f_c) + M(f + f_c)]$$

Coherent Detection



$$\begin{aligned}
 v(t) &= A'_c \cos(2\pi f_c t + \phi) s(t) \\
 &= A_c A'_c \cos(2\pi f_c t) \cos(2\pi f_c t + \phi) m(t) \\
 &= \frac{1}{2} A_c A'_c \cos(4\pi f_c t + \phi) m(t) + \frac{1}{2} A_c A'_c \cos \phi m(t)
 \end{aligned}$$

Coherent Detection



- Cut-off frequency $> W$ but less than $2f_c - W$

$$v_o(t) = \frac{1}{2} A_c A'_c \cos \phi m(t)$$

- Quadrature null effect of coherent detector? \gg perfect Synch.

Angle Modulation

$$s(t) = A_c \cos[\theta_i(t)]$$

$$f_{\Delta t}(t) = \frac{\theta_i(t + \Delta t) - \theta_i(t)}{2\pi \Delta t}$$

$$\begin{aligned} f_i(t) &= \lim_{\Delta t \rightarrow 0} f_{\Delta t}(t) \\ &= \lim_{\Delta t \rightarrow 0} \left[\frac{\theta_i(t + \Delta t) - \theta_i(t)}{2\pi \Delta t} \right] \\ &= \frac{1}{2\pi} \frac{d\theta_i(t)}{dt} \end{aligned}$$

UM

$$\theta_i(t) = 2\pi f_c t + \phi_c$$

PM

$$\theta_i(t) = 2\pi f_c t + k_p m(t)$$

$$s(t) = A_c \cos[2\pi f_c t + k_p m(t)]$$

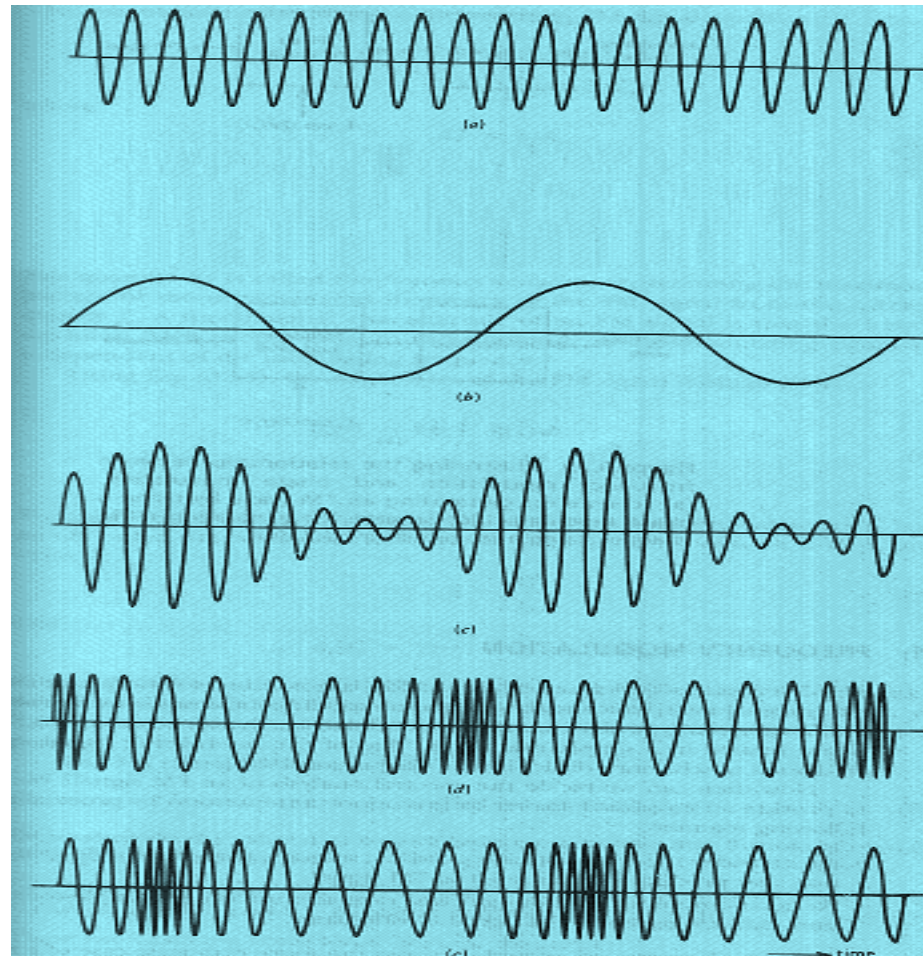
FM

$$f_i(t) = f_c + k_f m(t)$$

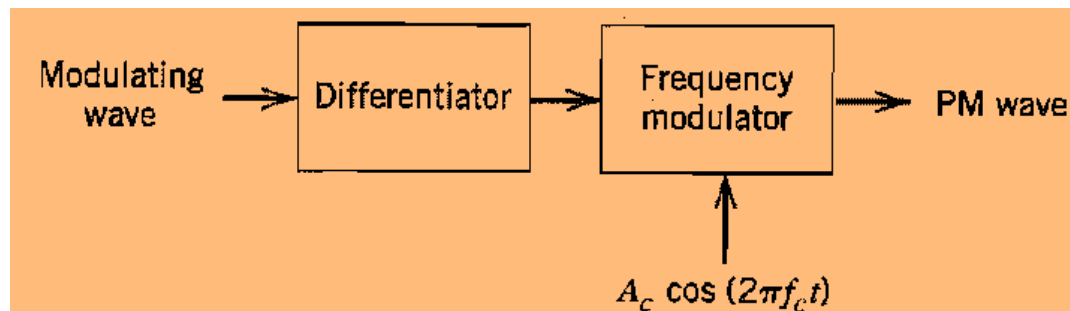
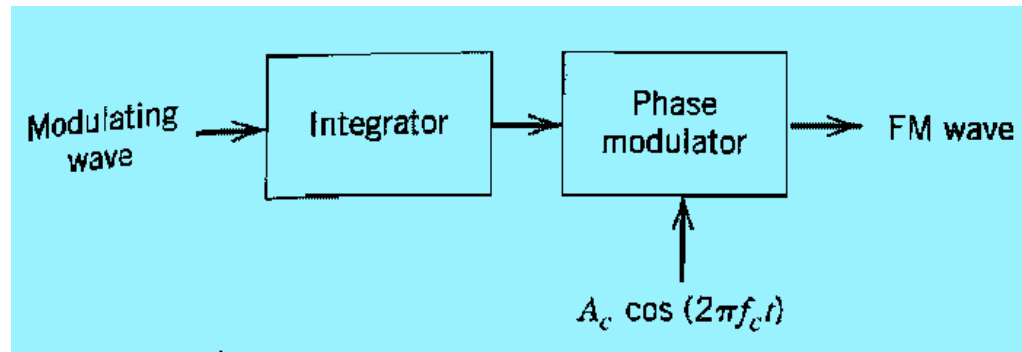
$$\theta_i(t) = 2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau$$

$$s(t) = A_c \cos \left[2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau \right]$$

Angle Modulation



Angle Modulation



Frequency Modulation

Spectral Analysis: Difficult compared to simple AM

$$m(t) = A_m \cos(2\pi f_m t)$$

The instantaneous frequency of the resulting FM signal equals

$$\begin{aligned} f_i(t) &= f_c + k_f A_m \cos(2\pi f_m t) \\ &= f_c + \Delta f \cos(2\pi f_m t) \end{aligned}$$

where

$$\Delta f = k_f A_m$$

Frequency deviation

Proportional to amplitude of $m(t)$ and independent of f_m

Frequency Modulation

$$\theta_i(t) = 2\pi \int_0^t f_i(\tau) d\tau$$

$$= 2\pi f_c t + \frac{\Delta f}{f_m} \sin(2\pi f_m t)$$

Modulation index



$$\beta = \frac{\Delta f}{f_m}$$

Phase deviation

$$\theta_i(t) = 2\pi f_c t + \beta \sin(2\pi f_m t)$$

$$s(t) = A_c \cos[2\pi f_c t + \beta \sin(2\pi f_m t)]$$

NB or WB

Frequency Modulation

NB Frequency Modulation

$$s(t) = A_c \cos(2\pi f_c t) \cos[\beta \sin(2\pi f_m t)] - A_c \sin(2\pi f_c t) \sin[\beta \sin(2\pi f_m t)]$$

and

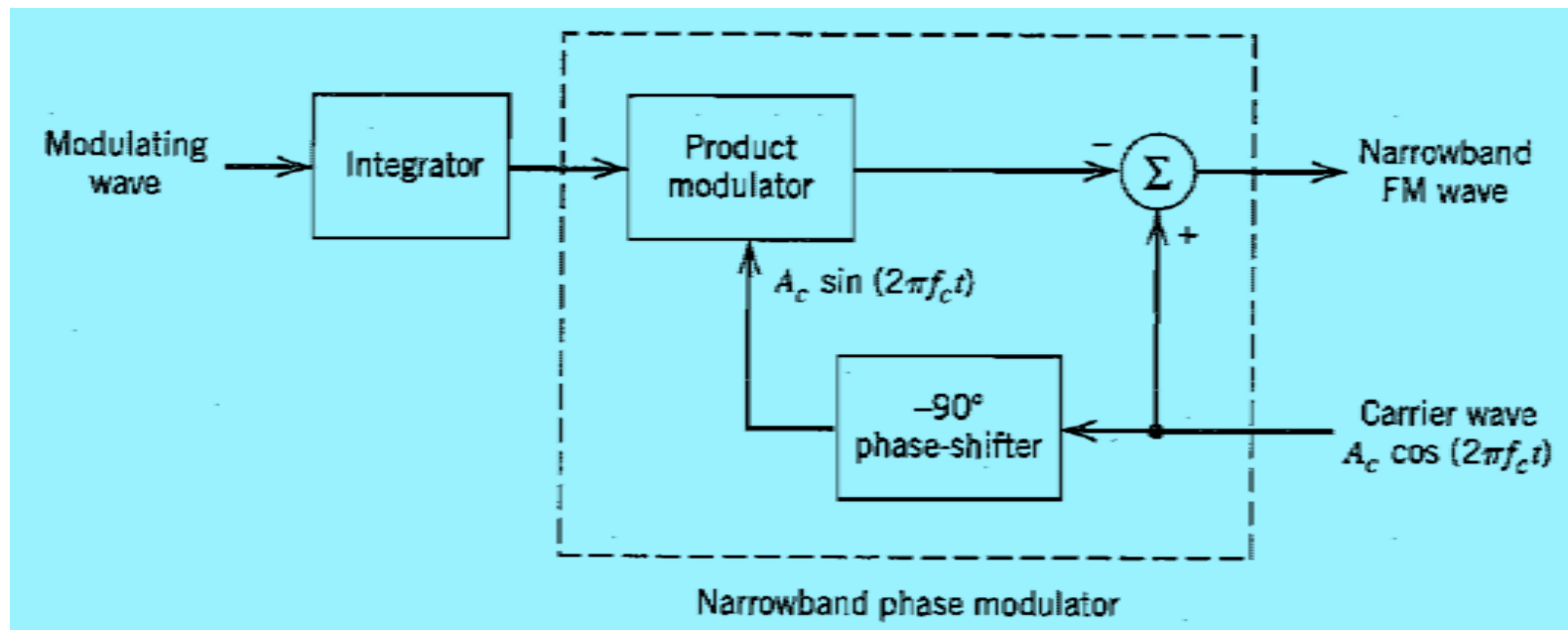
$$\cos[\beta \sin(2\pi f_m t)] \approx 1$$

$$\sin[\beta \sin(2\pi f_m t)] \approx \beta \sin(2\pi f_m t)$$

Hence, Equation (2.34) simplifies to

$$s(t) \approx A_c \cos(2\pi f_c t) - \beta A_c \sin(2\pi f_c t) \sin(2\pi f_m t)$$

Frequency Modulation



1. The envelope contains a residual amplitude modulation and, therefore, varies with time.
2. For a sinusoidal modulating wave, the angle $\theta_i(t)$ contains harmonic distortion in the form of third- and higher-order harmonics of the modulation frequency f_m .

Modulation index < 0.3 rad

Frequency Modulation

$$s(t) \approx A_c \cos(2\pi f_c t) + \frac{1}{2} \beta A_c \{\cos[2\pi(f_c + f_m)t] - \cos[2\pi(f_c - f_m)t]\}$$

$$s_{AM}(t) = A_c \cos(2\pi f_c t) + \frac{1}{2} \mu A_c \{\cos[2\pi(f_c + f_m)t] + \cos[2\pi(f_c - f_m)t]\}$$

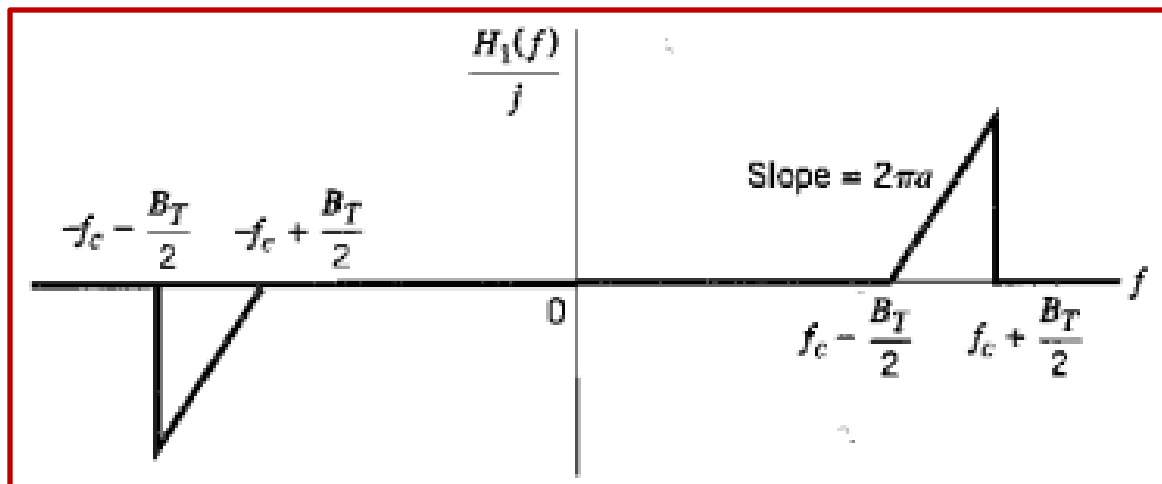
Difference is algebraic sign of LSB is reversed,
so the Tx BW is the same as that of AM ($2f_m$)

Demodulation of FM Signals

Using Frequency Discriminator

Slope Circuit followed by Envelope detector

$$H_1(f) = \begin{cases} j2\pi a \left(f - f_c + \frac{B_T}{2} \right), & f_c - \frac{B_T}{2} \leq f \leq f_c + \frac{B_T}{2} \\ j2\pi a \left(f + f_c - \frac{B_T}{2} \right), & -f_c - \frac{B_T}{2} \leq f \leq -f_c + \frac{B_T}{2} \\ 0, & \text{elsewhere} \end{cases}$$



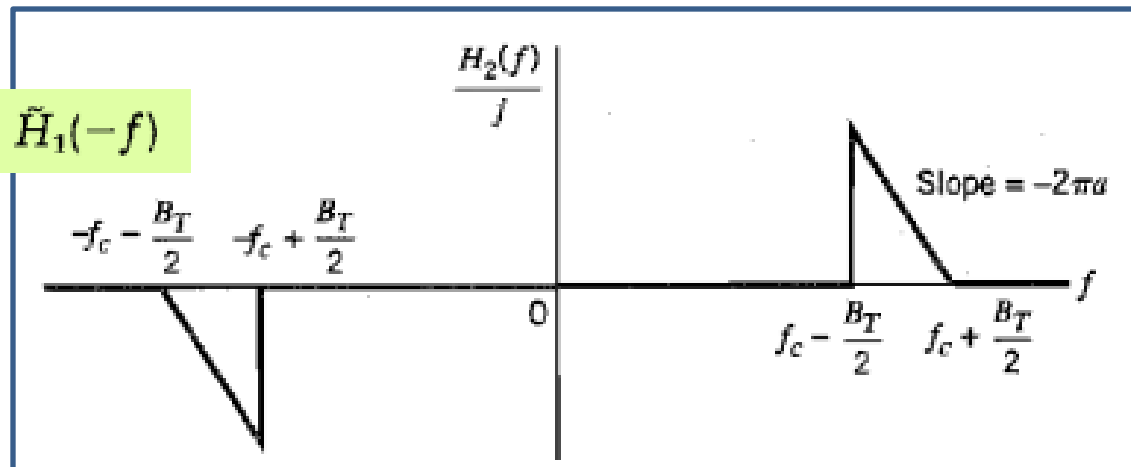
Demodulation of FM Signals

$$\begin{aligned}
 s_1(t) &= \text{Re}[\tilde{s}_1(t) \exp(j2\pi f_c t)] \\
 &= \pi B_T a A_c \left[1 + \frac{2k_f}{B_T} m(t) \right] \cos \left[2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau + \frac{\pi}{2} \right]
 \end{aligned}$$

$$|\tilde{s}_1(t)| = \pi B_T a A_c \left[1 + \frac{2k_f}{B_T} m(t) \right]$$

$$\left| \frac{2k_f}{B_T} m(t) \right| < 1 \quad \text{for all } t$$

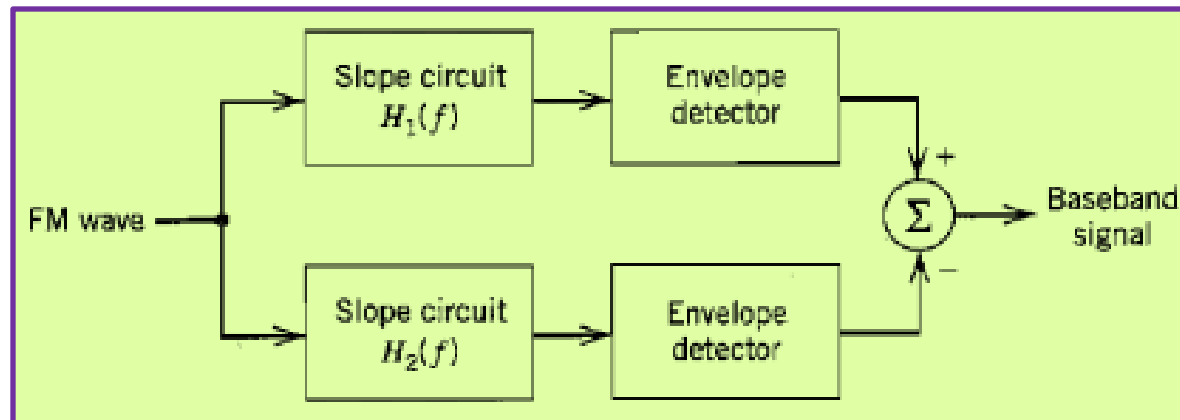
$$\tilde{H}_2(f) = \tilde{H}_1(-f)$$



Demodulation of FM Signals

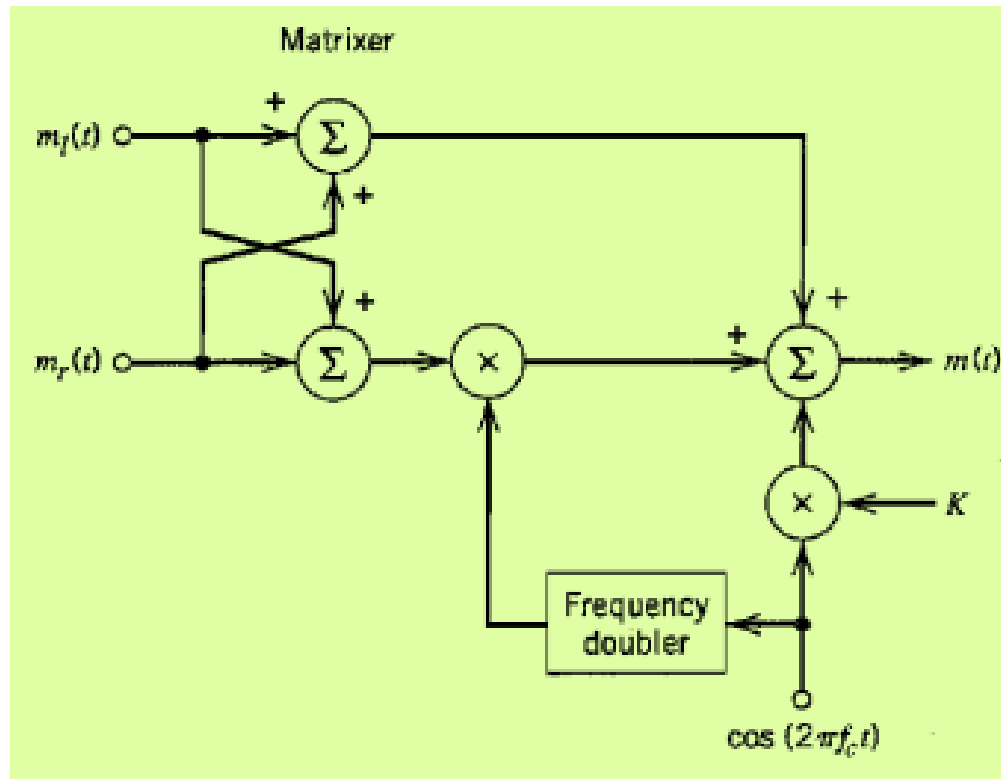
$$|\tilde{s}_2(t)| = \pi B_T a A_c \left[1 - \frac{2k_f}{B_T} m(t) \right]$$

$$\begin{aligned} s_o(t) &= |\tilde{s}_1(t)| - |\tilde{s}_2(t)| \\ &= 4\pi k_f a A_c m(t) \end{aligned}$$

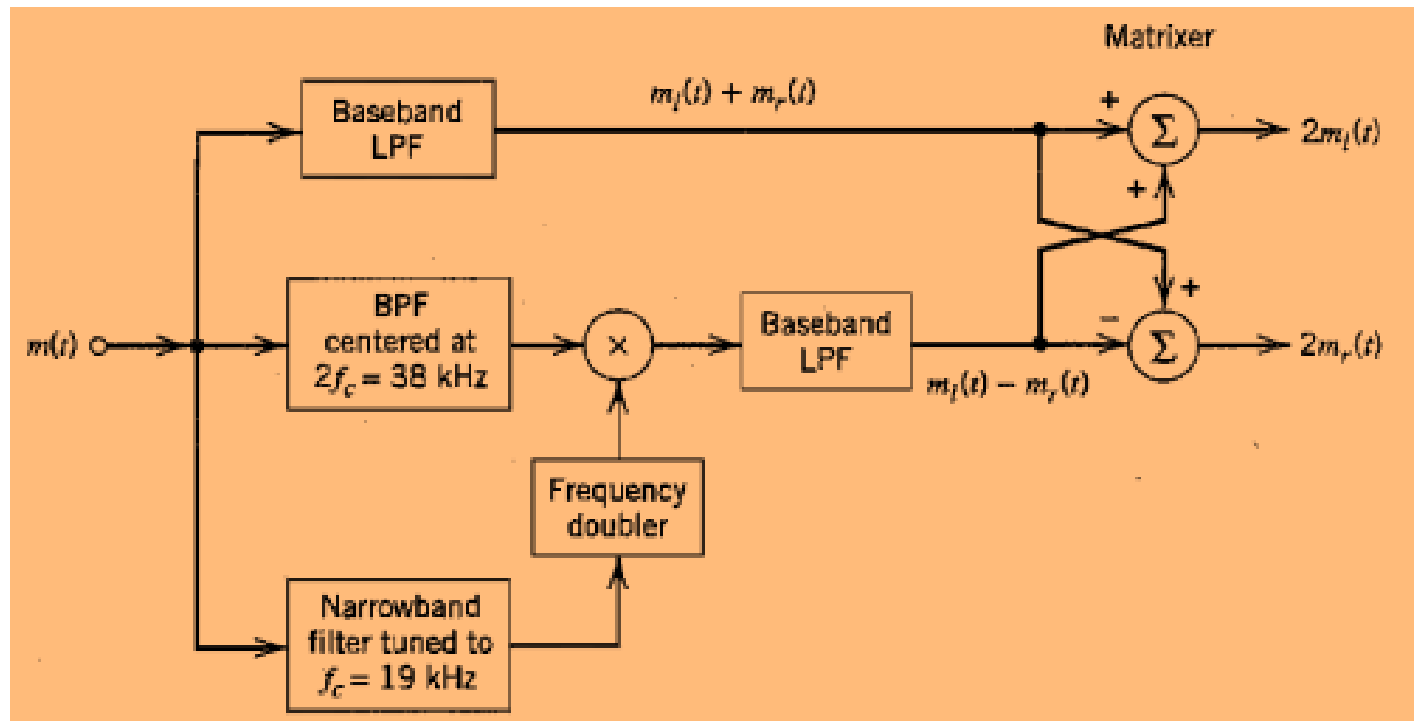


Balanced Frequency Discriminator

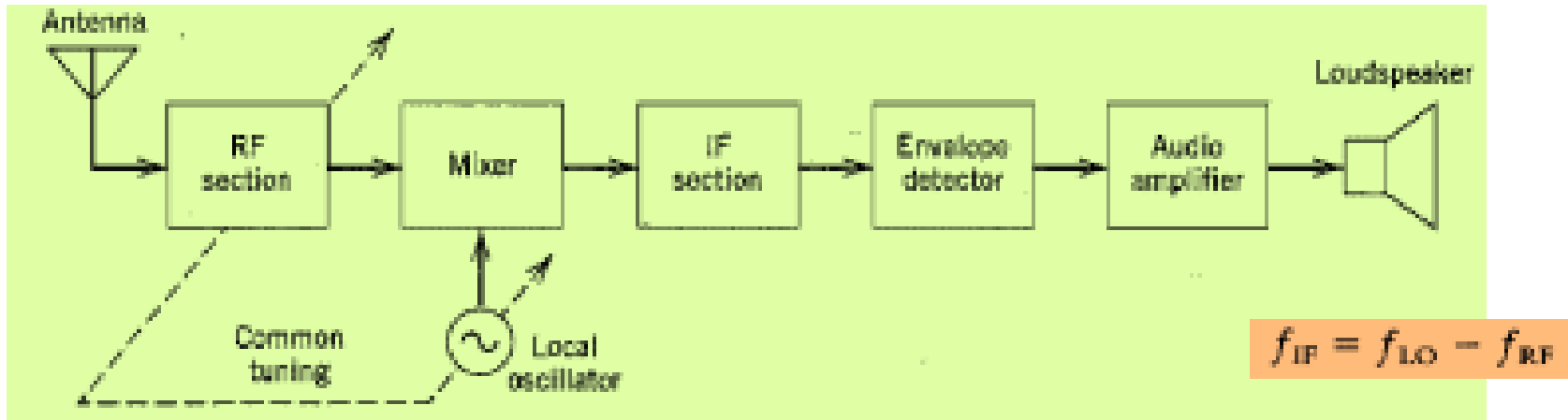
FM Stereo Multiplexing



FM Stereo Multiplexing



Superheterodyne Receiver



Disadvantage: Image Frequency

Q & A

